

Wireless Innovation Without Bounds



Challenges and Design Considerations for Integrated VCOs in Wireless Communications

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Outline

- ❑ Introduction – Trends
- ❑ Basics – LC VCOs
- ❑ Issues and trade-offs of integrated VCOs
 - ❑ Will **not** cover various methods of low phase noise design
 - ❑ Will cover trade-offs that are not often covered in the literature.
- ❑ Modern fine-line processes: good or bad?
- ❑ Conclusions

Introduction

- ❑ Assumptions for this presentation: Integrated VCOs in ...
 - ❑ Wireless market
 - ❑ High volume market (cellular phone, tuners, WLANs, etc.)
 - ❑ LC VCOs

- ❑ SoC trend: High level of integration
 - ❑ Multiple standards in one chip (e. g., EDGE, WCDMA, LTE, etc.)
 - ❑ Widely different frequency ranges
 - ❑ Different requirements: low noise, low power, etc.
 - ❑ Multiple VCOs in a chip along with multiple LNAs, mixers, PAs, baseband blocks, micro-controller, etc.

- ❑ Process trends: CMOS
 - ❑ 0.18 μm \rightarrow 0.13 μm \rightarrow 90 nm \rightarrow 65 nm \rightarrow ...



Introduction – Underlying Theme

- ❑ “RFIC” design requires “analog” IC design skills.
- ❑ Modern CMOS process → many optional devices → more tools for engineers to innovate.

Basics

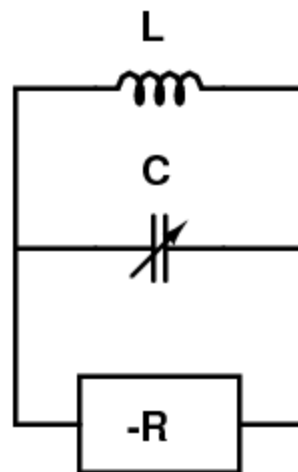
Basic Structures for LC VCO

Design Variables

Inductance

Variable capacitance

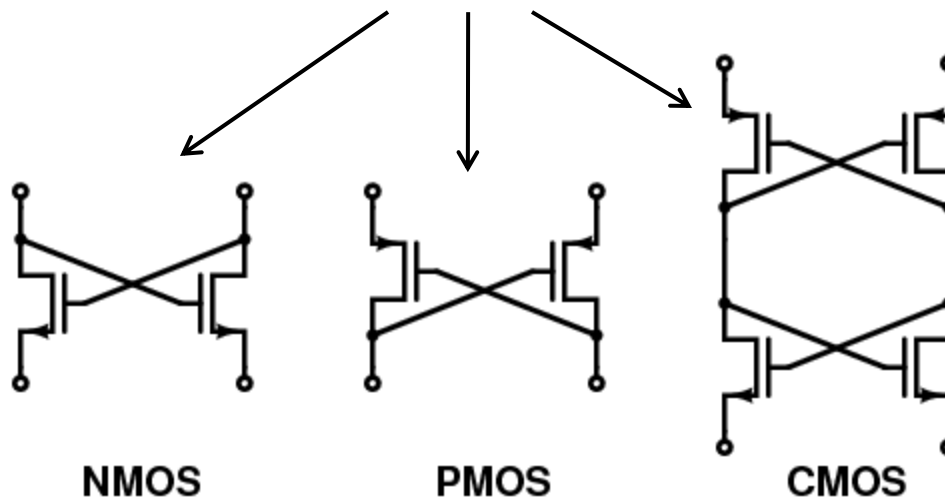
Negative resistance



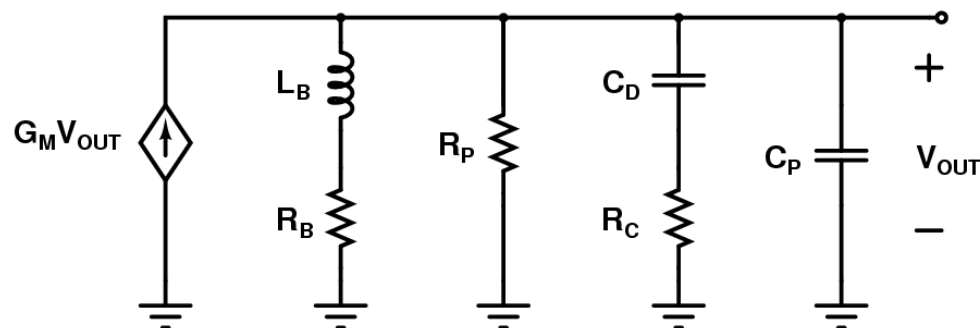
Normally fixed

MIM, MOS, MOM, PNJ varactor, etc.

Change in sizes or current
Choice of I/O or core devices



Basics: Conditions for Oscillations



Conditions for oscillations

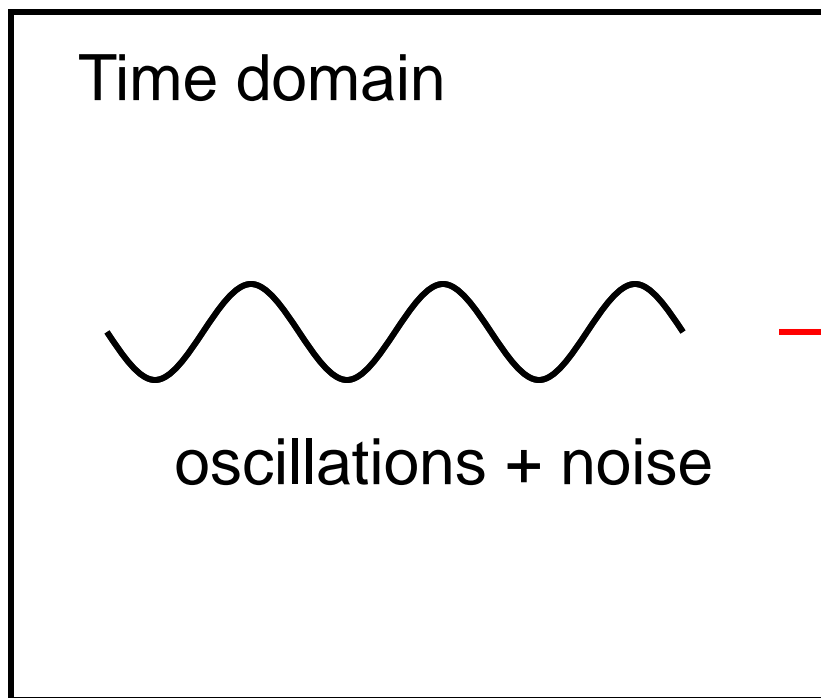
$$\omega_0^2 \approx \frac{1}{L_B \cdot (C_D + C_P)}$$

$$L(\omega_0) = G_M \cdot (Q_B^2 R_B \parallel R_P \parallel Q_C^2 R_C) \geq 1$$

This needs to be satisfied for PVT (process, voltage, temperature) variations with a margin !

Basics: Phase Noise

Phase Noise = Equivalent of a "Signal to Noise Ratio" for an Oscillator

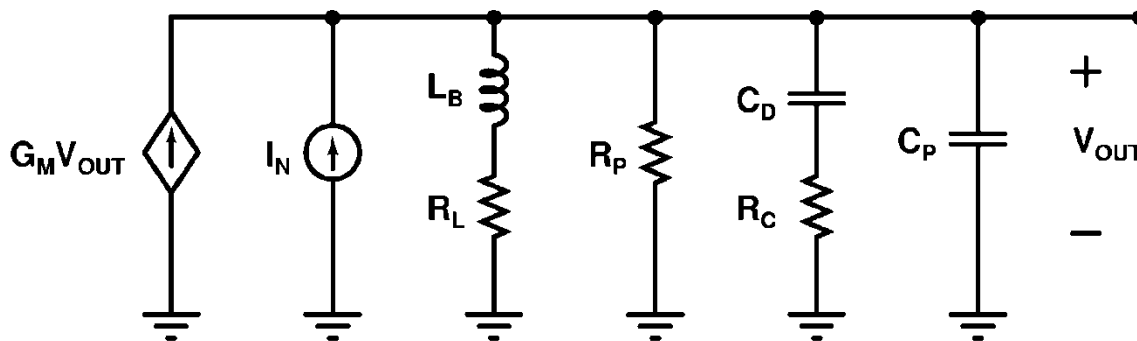
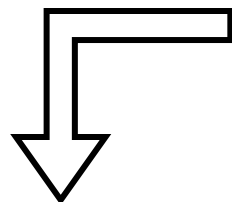


Why is phase noise important ?

- Phase noise can affect receiver sensitivity (i.e. blocker + LO noise mix onto desired channel)

Basics: Phase Noise (linear model)

$$C_P = \gamma C_D$$



$$S_{\theta}(\delta\omega) = \frac{\frac{k_b T}{2} \times (1 + F_{GM}) \times \left(\frac{1}{\omega_0^2 (1 + \gamma)^2 C_D^2 R_P} + R_L + \frac{R_C}{(1 + \gamma)^2} \right) \times \left(\frac{\omega_0}{\delta\omega} \right)^2}{\frac{V_{out,peak}^2}{2}}$$

G_M noise loss in LC tank -20 dB/dec
signal swing

How to get good phase noise?

- ❑ Use **high-Q devices** if available
 - ❑ Thick metal for high-Q on-chip inductors
 - ❑ High-Q PN-junction and MOS varactors

- ❑ Very **careful layout !!**
 - ❑ High-Q layout vs. low parasitic capacitance layout
 - ❑ Try isolating the control lines

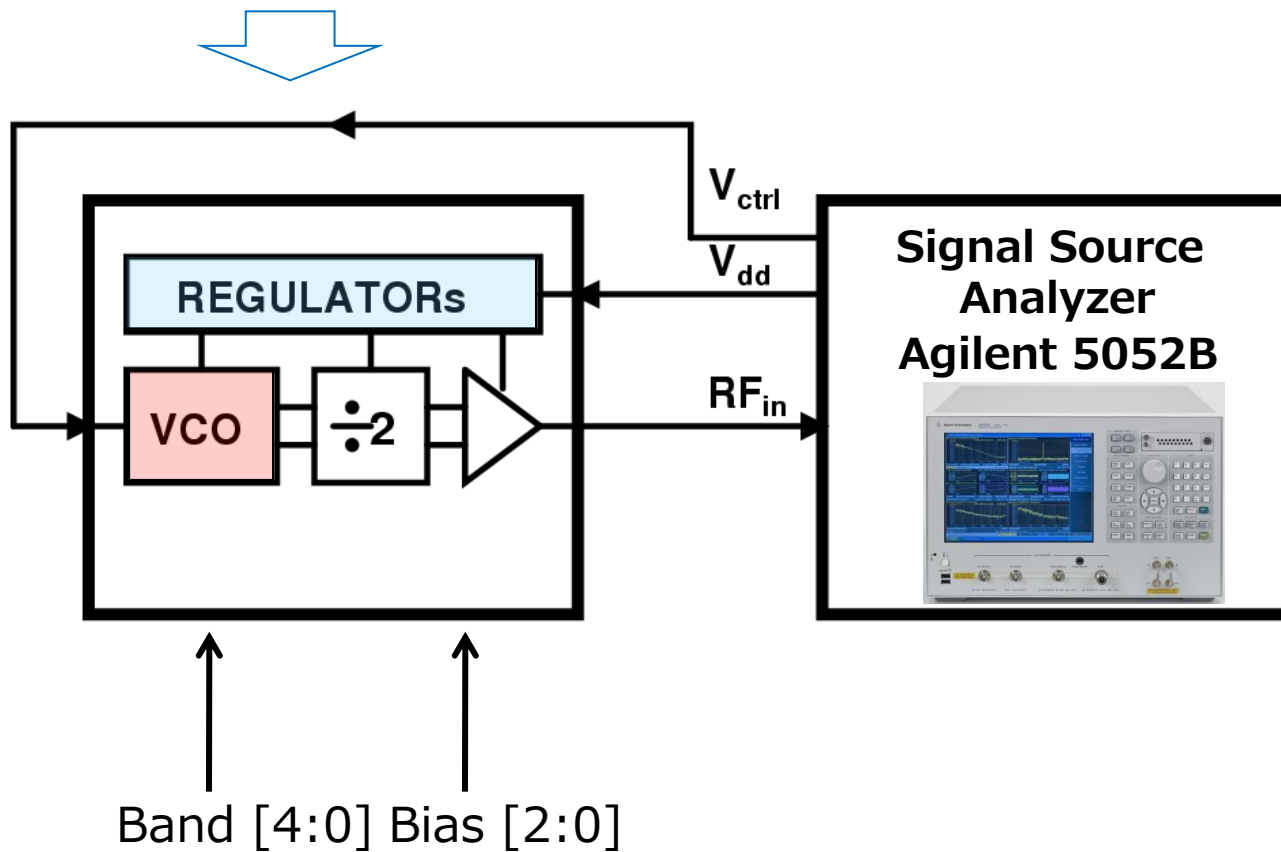
- ❑ **Maximize** oscillator voltage **swing**

- ❑ Care with the oscillation amplitude
 - ❑ Voltage limiting vs. current limiting

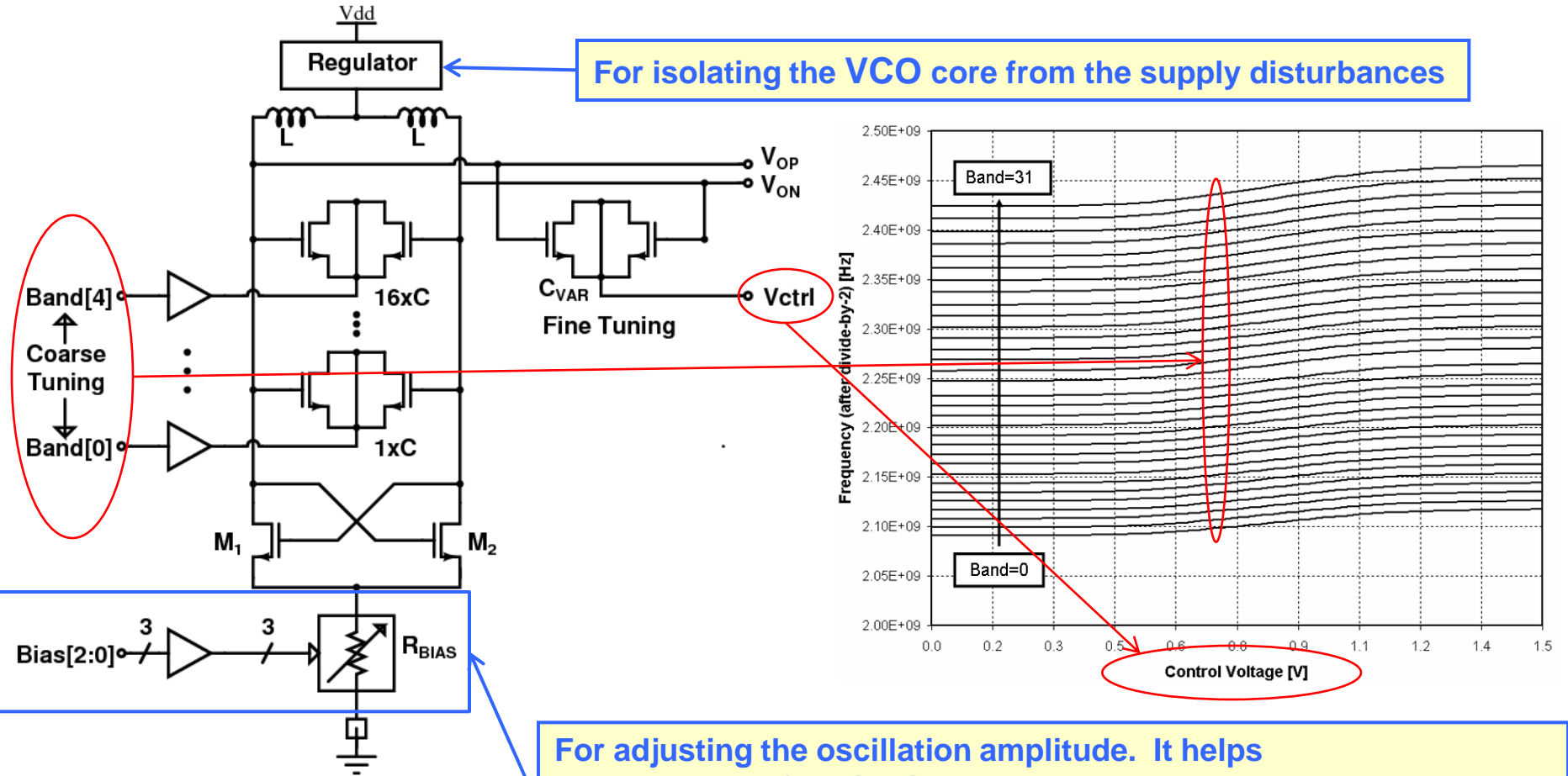
Case Study

Measurement Setup

A part of a large SoC chip



A Realistic VCO: A Simplified Regulated NMOS VCO

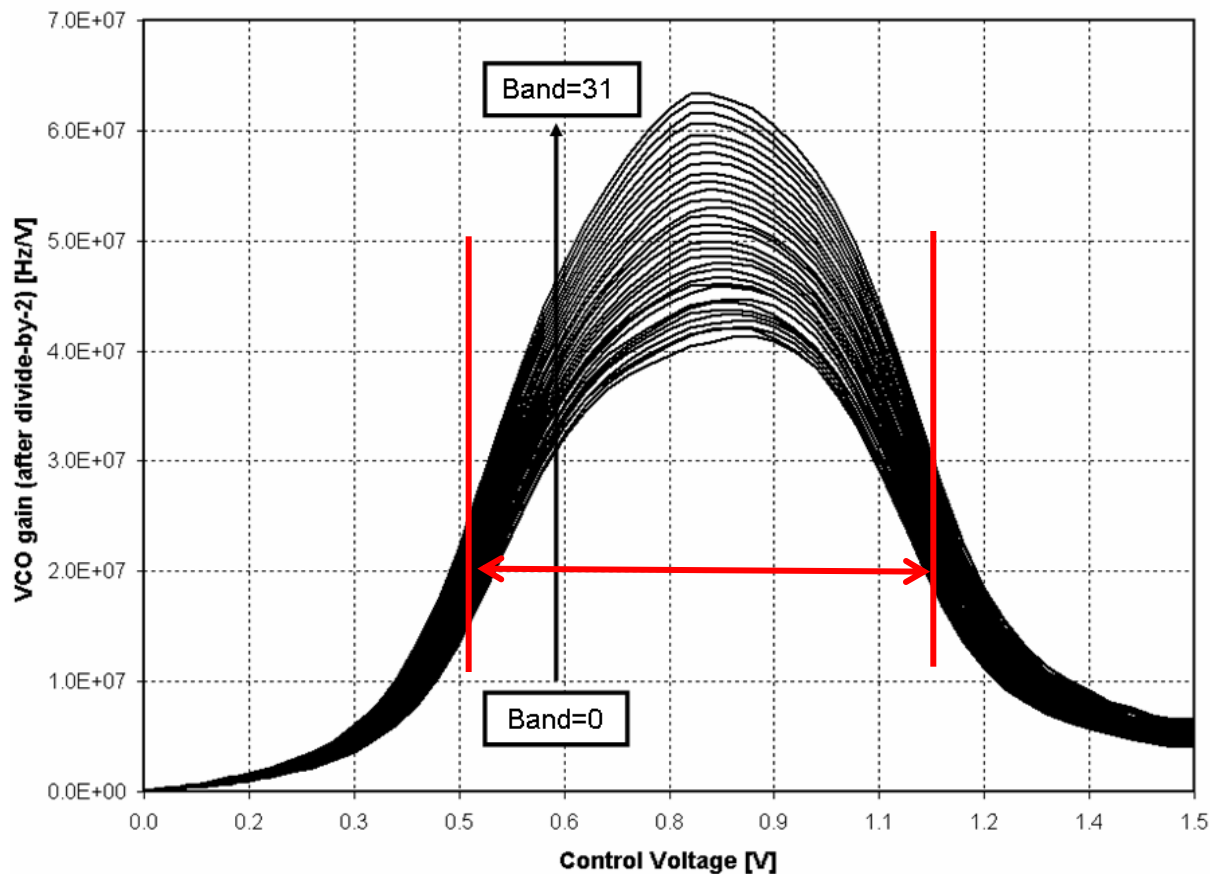


For adjusting the oscillation amplitude. It helps

- Robustness of oscillation
- Phase noise.

Use of resistors instead of current source → generally lower noise

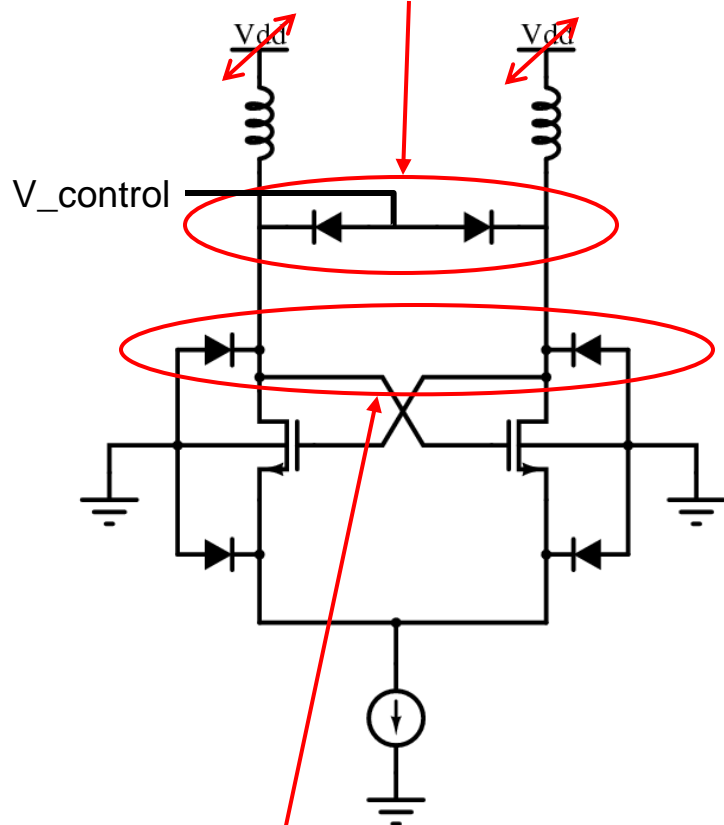
A Realistic VCO: Measured VCO Gain



VCO Gain @ 0.8V = 41MHz/V to 63MHz/V

Issue #1: Supply Pushing

Example: Varactors connected to Vcc



Example: Transistor diffusion diodes connect to Vdd.

Supply Pushing = Oscillation Frequency Sensitivity to Supply Voltage Variation

$$\text{Supply Pushing} = \frac{df_0}{dV_{CC}} [\text{Hz} / \text{V}]$$

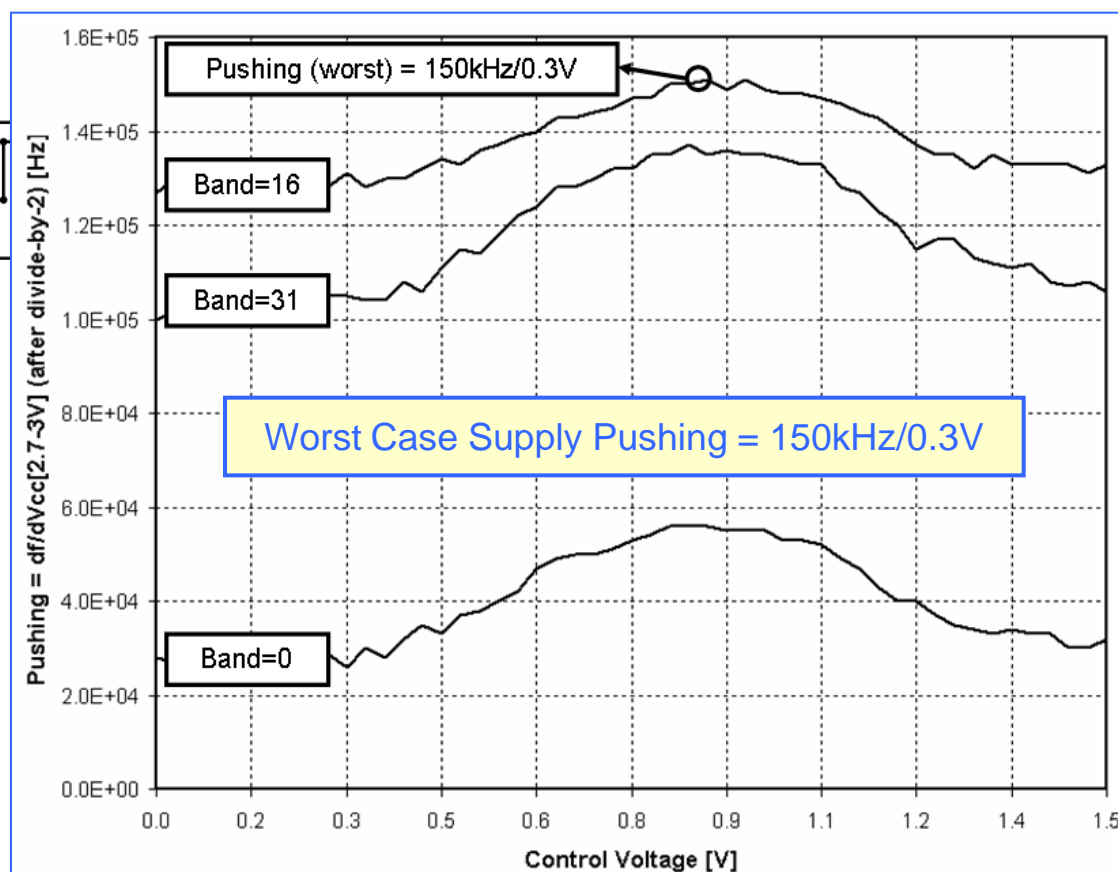
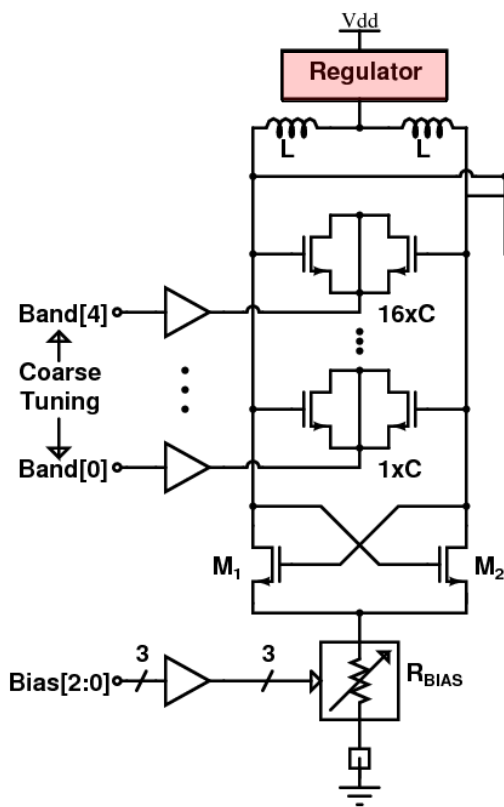
Why is pushing important ?

- PLLs using VCOs with high pushing can un-lock when various parts of the system get enabled or disabled

Possible solutions ?

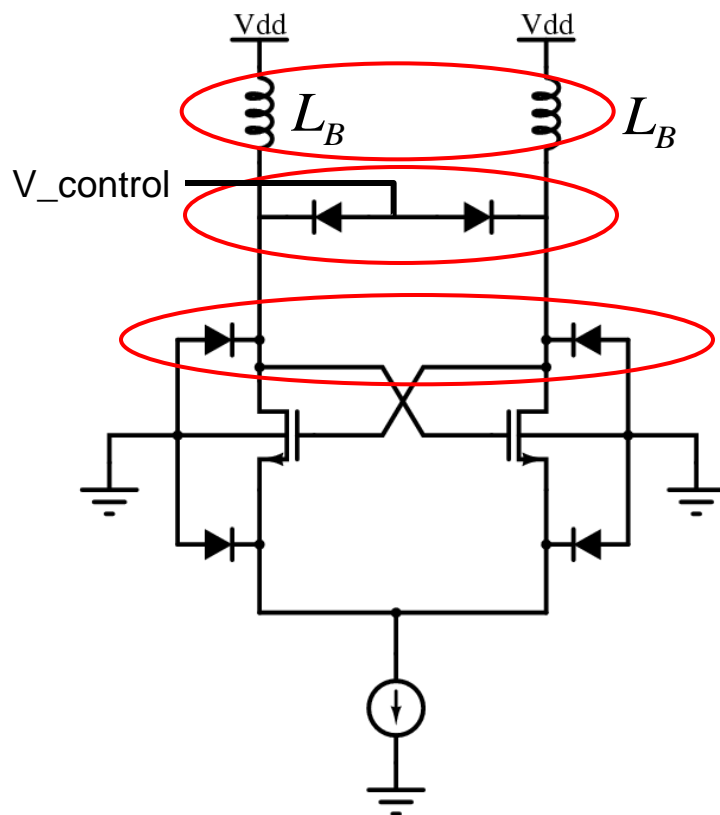
- PMOS based → less sensitive to V_{DD} variations
- Regulate the supply (V_{DD})

Issue #1: Supply Pushing: A Solution → Regulate Vdd



Issue #2: Temperature Variations

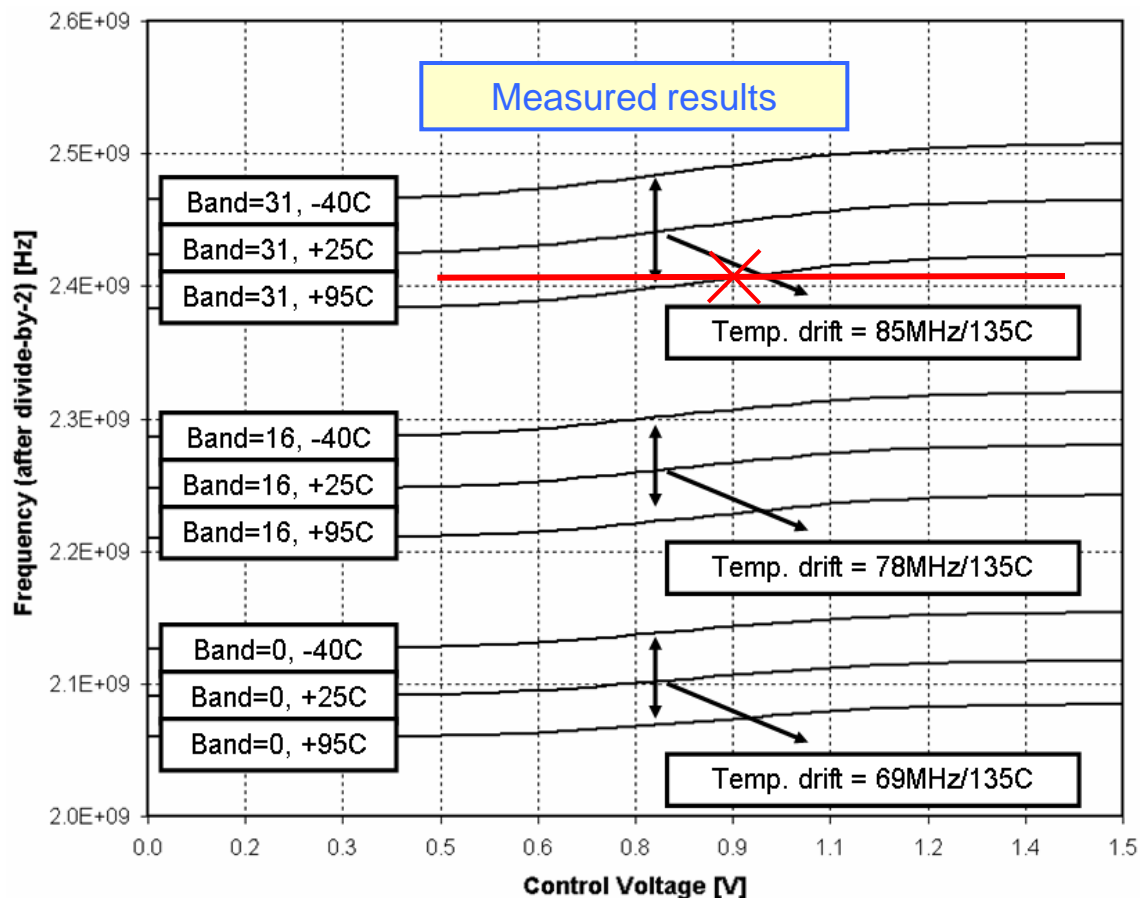
An Example



- Temp \uparrow \rightarrow the “capacitance” \uparrow
- When Q_B is small, the variation of Q_B affects the effective inductance value (L'_B)
 - Thus, the temperature dependence of the loss affects greatly.

$$L'_B = L_B \left(1 + \frac{1}{Q_B^2} \right)$$

Issue #2: Temperature Variations

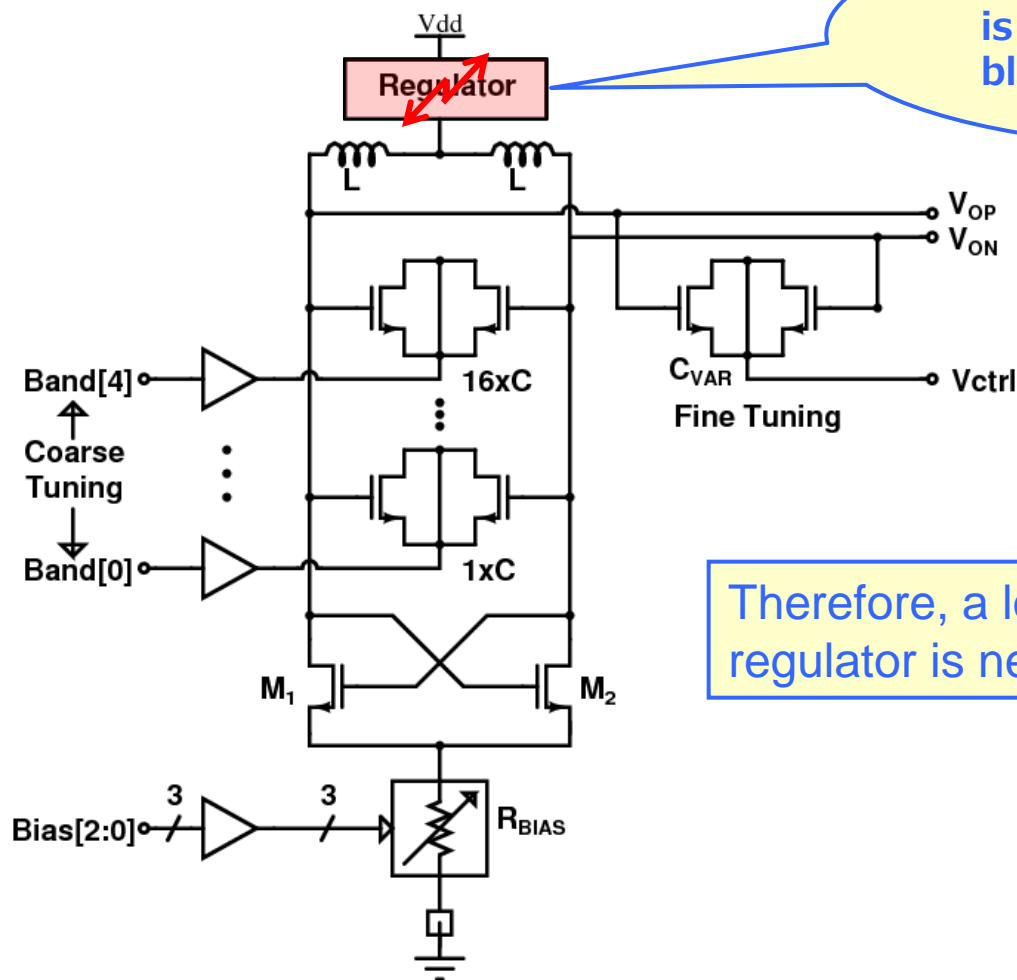


- ❑ Temperature behavior is difficult to simulate accurately.
- ❑ Need to model the temp. coefficient of inductor losses !

Issue #2: Possible Solutions

- ❑ Purposely **embed a temperature dependence to the VCO control** to reduce the temperature dependence.
- ❑ **Increase K_v** (VCO gain) to cover the temperature variations.
- ❑ **Sense temperature change and adjust a fine-tuning varactor** to negate the temperature variations.
- ❑ Do nothing if the system allows frequent auto-band calibration (for example, TDMA-type system)

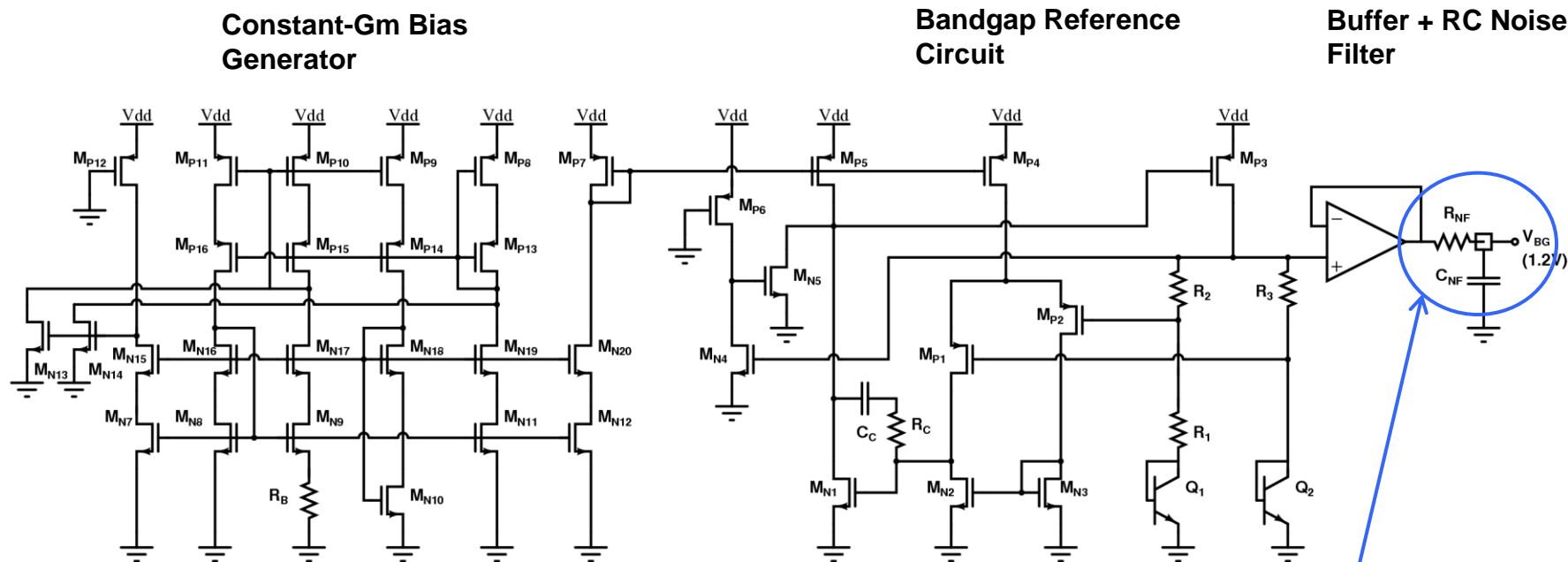
Issue #3: Regulator vs Phase Noise



Bandgap based regulator is often used. But, this block generates noise!!

Therefore, a low-noise bandgap based regulator is needed.

Issue #3: An Example Bandgap Circuit

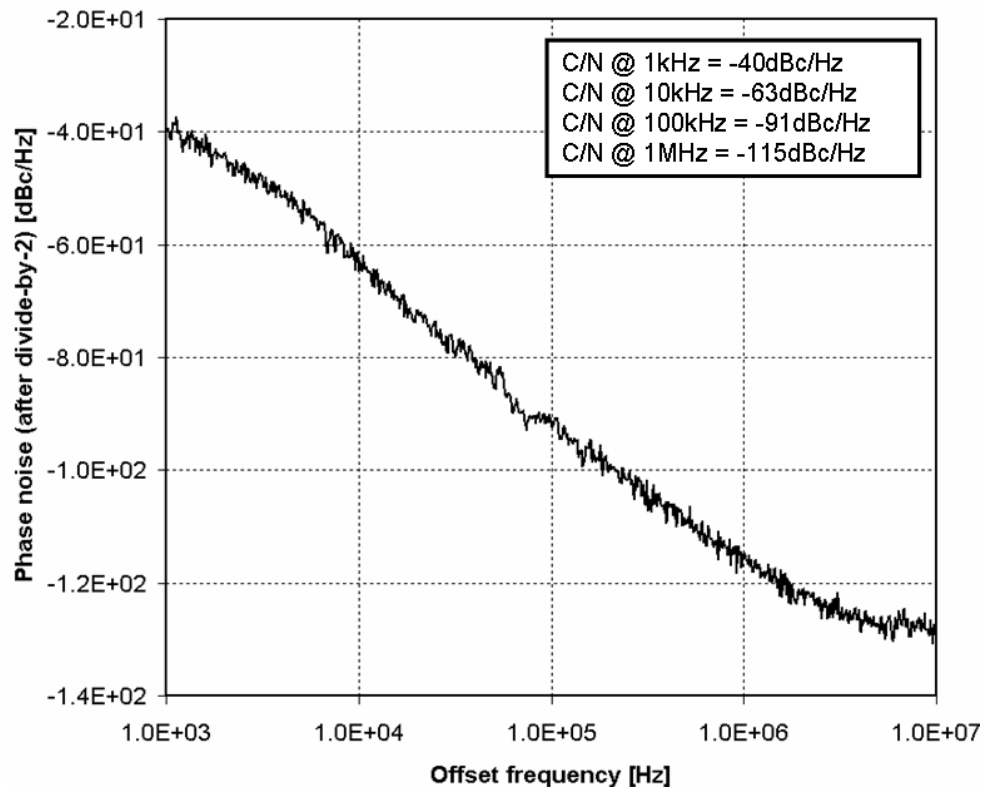


Noise filter to reduce bias noise impact on VCO phase noise

A possible problem → Slow turn-on → May need a speed-up circuit.

Analog IC design skill becomes essential.

Issue #3: Measured Phase Noise



Phase Noise @ 1MHz = -115dBc/Hz

Modern Fine-Line
Processes
Good or Bad?

VCO Design: Modern processes: good or bad?

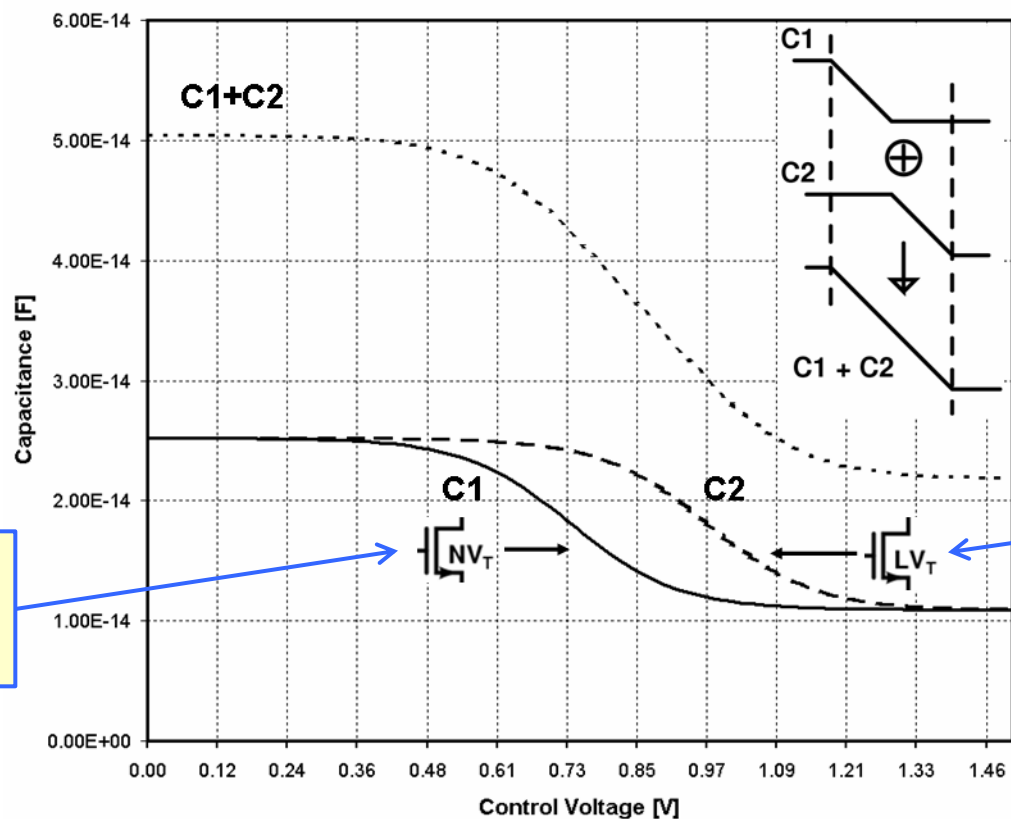
Cons

- Break-down voltage gets smaller but oscillation amplitude should be large for good phase noise.
- $1/f$ noise is often larger.

Pros

- Faster devices.
- Tuning range increases.
- A variety of devices such as native devices, devices with different threshold (normal, low, etc.) are available.
 - More “tools” for engineers to innovate !

Issue #4: Linear Range Extension: An Example Use of Device Options

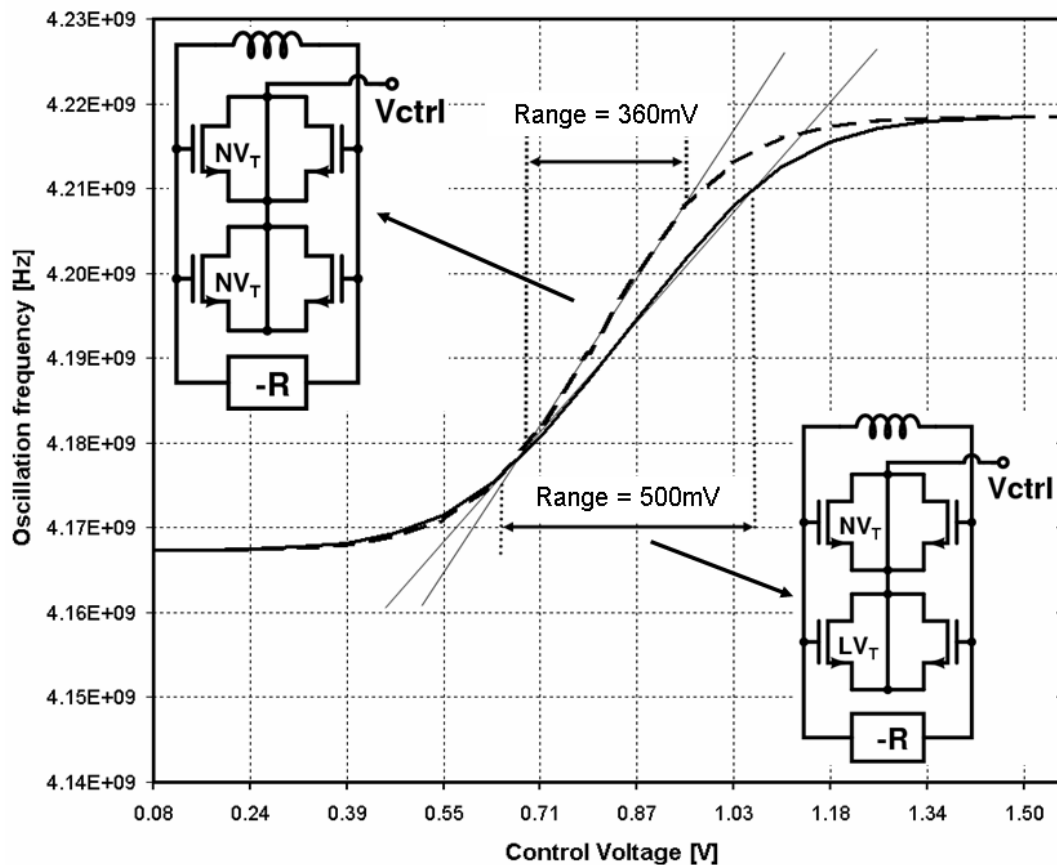


Normal
Threshold
Device

Low Threshold
Device

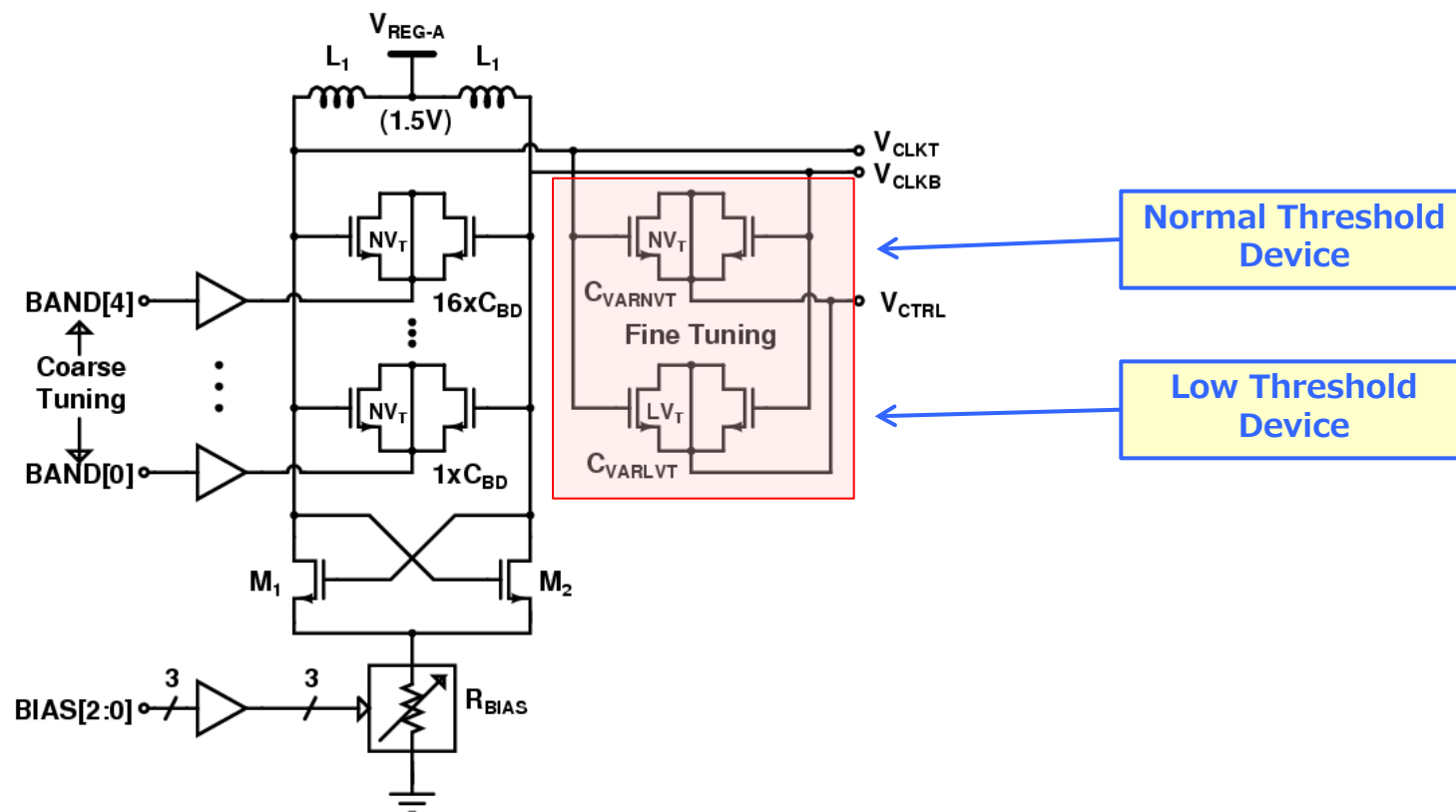
Possible to linearize MOS capacitance by superposition of normal & low threshold transistors

Simulated Range Extension



Linear tuning range improvement of 38%

VCO Schematic using the Range Extension



Conclusions

- ❑ Modern transceiver ICs require multiple VCOs integrated on chip.
- ❑ Complex trade-offs exist among oscillation frequency, K_v (VCO gain), required frequency coverage, supply pushing, phase noise, power, area (\$), PVT variations, startup time, etc.
 - ❑ The trade-offs get more complicated as the process shrinks.
- ❑ Modern processes provide various options for active and passive devices → More “tools” for engineers.

- ❑ Therefore, more room for engineers to innovate !

Acknowledgment

- The author would like to thank Aleksander Dec and Hiroshi Akima for useful discussions.